## Advanced Systems Lab

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Lecture: Discrete Fourier transform, fast Fourier transforms

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## Linear Transforms

Overview: Transforms and algorithms
Discrete Fourier transform
Fast Fourier transforms

After that:

- Optimized implementation and autotuning (FFTW)
- Automatic program synthesis (Spiral)


## FFT References

Complexity: Bürgisser, Clausen, Shokrollahi, Algebraic Complexity Theory, Springer, 1997
History: Heideman, Johnson, Burrus: Gauss and the History of the Fast Fourier Transform, Arch. Hist. Sc. 34(3) 1985

FFTs:

- Cooley and Tukey, An algorithm for the machine calculation of complex Fourier series," Math. of Computation, vol. 19, pp. 297-301, 1965
- Nussbaumer, Fast Fourier Transform and Convolution Algorithms, 2nd ed., Springer, 1982
- van Loan, Computational Frameworks for the Fast Fourier Transform, SIAM, 1992
- Tolimieri, An, Lu, Algorithms for Discrete Fourier Transforms and Convolution, Springer, 2nd edition, 1997
- Franchetti, Püschel, Voronenko, Chellappa and Moura, Discrete Fourier Transform on Multicore, IEEE Signal Processing Magazine, special issue on "Signal Processing on Platforms with Multiple Cores", Vol. 26, No. 6, pp. 90-102, 2009


## Linear Transforms

Very important class of functions: signal processing, communication, scientific computing, ...

Mathematically: Change of basis = Multiplication by a fixed matrix $T$

$$
\begin{aligned}
& \left(\begin{array}{c}
y_{0} \\
y_{1} \\
\vdots \\
y_{n-1}
\end{array}\right)=y=T x \quad x=\left(\begin{array}{c}
x_{0} \\
x_{1} \\
\vdots \\
x_{n-1}
\end{array}\right) \\
& \text { Output } \\
& T=\left[t_{k, \ell}\right]_{0 \leq k, \ell<n} \\
& \text { Input }
\end{aligned}
$$

Equivalent definition: Summation form

$$
y_{k}=\sum_{\ell=0}^{n-1} t_{k, \ell} x_{\ell}, \quad 0 \leq k<n
$$

## Linear Transforms

Compute: $\mathrm{y}=\mathrm{Tx} \quad \mathrm{x}$ : input vector, y : output vector, T : fixed transform matrix
Example: Discrete Fourier transform (DFT)

1. form (standard in signal processing):

$$
\begin{aligned}
\text { given: } x_{0}, \ldots, x_{n-1} \\
\text { compute: } \quad \begin{aligned}
y_{k} & =\sum_{\ell=0}^{n-1} e^{-2 k \ell \pi i / n} x_{\ell}, \quad k=0, \ldots, n-1 \\
& =\sum_{\ell=0}^{n-1} \omega_{n}^{k \ell} x_{\ell}, \quad k=0, \ldots, n-1, \quad \omega_{n}=e^{-2 \pi i / n}
\end{aligned} . \quad \text { primitive nth root of } 1
\end{aligned}
$$

2. form (we will use):
given: $\left(x_{0}, \ldots, x_{n-1}\right)^{T}$
compute: $\quad y=\mathbf{D F T}_{n} \cdot x, \quad \mathbf{D F T}_{n}=\left[\omega_{n}^{k \ell}\right]_{0 \leq k, \ell<n}$

How does the $\mathrm{DFT}_{2}$ matrix look?
Second row of $\mathrm{DFT}_{4}$ matrix?

## Smallest Relevant Example: DFT, Size 2

Transform (matrix): $\quad \mathbf{D F T}_{2}=\left[\begin{array}{rr}1 & 1 \\ 1 & -1\end{array}\right] \quad \begin{aligned} & \text { How many ops to compute } \\ & \text { the } D F T_{2} \text { of a vector? }\end{aligned}$
(the $\mathrm{DFT}_{2}$ of a vector?

Computation: $\quad y=\left[\begin{array}{rr}1 & 1 \\ 1 & -1\end{array}\right] x \quad$ or $\quad \begin{aligned} & y_{0}=x_{0}+x_{1} \\ & y_{1}=x_{0}-x_{1}\end{aligned}$

As graph (direct acyclic graph or DAG):

called a butterfly


## DFT, Size 4

$\mathrm{DFT}_{4}=\left[\begin{array}{rrrr}1 & 1 & 1 & 1 \\ 1 & i & -1 & -i \\ 1 & -1 & 1 & -1 \\ 1 & -i & -1 & i\end{array}\right]$

How many (complex) operations to compute the $\mathrm{DFT}_{4}$ of a (complex) vector?
$y=\mathrm{DFT}_{4} \cdot x$

## Transforms: Examples

A few dozen transforms are relevant
Some examples

$$
\begin{array}{rll}
\text { DFT }_{n} & =\left[e^{-2 k \ell \pi i / n}\right]_{0 \leq k, \ell<n} \\
\mathbf{R D F T}_{n} & =\left[r_{k \ell}\right]_{0 \leq k, \ell<n}, \quad r_{k \ell}= \begin{cases}\cos \frac{2 \pi k \ell}{n}, & k \leq\left\lfloor\frac{n}{2}\right\rfloor \\
-\sin \frac{2 \pi k \ell}{n}, & k>\left\lfloor\frac{n}{2}\right\rfloor\end{cases} & \text { universal tool } \\
\text { DHT } & =\left[\begin{array}{ll}
\cos (2 k \ell \pi / n)+\sin (2 k \ell \pi / n)
\end{array}\right]_{0 \leq k, \ell<n} & \\
\mathbf{W H T}_{n} & =\left[\begin{array}{ll}
\mathbf{W H T}_{n / 2} & \mathbf{W H T}_{n / 2} \\
\mathbf{W H T}_{n / 2} & -\mathbf{W H T}_{n / 2}
\end{array}\right], \quad \mathbf{W H T}_{2}=\text { DFT }_{2} & \\
\mathbf{I M D C T}_{n} & =[\cos ((2 k+1)(2 \ell+1+n) \pi / 4 n)]_{0 \leq k<2 n, 0 \leq \ell<n} & \text { MPEG } \\
\text { DCT-2 }_{n} & =[\cos (k(2 \ell+1) \pi / 2 n)]_{0 \leq k, \ell<n} & \text { JPEG } \\
\text { DCT-3 }_{n} & =\mathbf{D C T - 2}_{n}^{T} & (\operatorname{transpose)} \\
\text { DCT-4 }_{n} & =[\cos ((2 k+1)(2 \ell+1) \pi / 4 n)]_{0 \leq k, \ell<n} &
\end{array}
$$

## Transform Algorithms

An algorithm for $\mathrm{y}=\mathrm{Tx}$ is given by a factorization

$$
T=T_{1} T_{2} \cdots T_{m}
$$

Namely, instead of $y=T x$ we can compute in steps
$t_{1}=T_{m} x$
$t_{2}=T_{m-1} t_{1}$
This reduces the op count only if:

- the $\mathrm{T}_{\mathrm{i}}$ are sparse
- $m$ is not too large


## Example: Cooley-Tukey Fast Fourier Transform (FFT), size 4



## Cooley-Tukey FFT, $\mathrm{n}=4$

Fast Fourier transform (FFT)

$$
\left[\begin{array}{rrrr}
1 & 1 & 1 & 1 \\
1 & i & -1 & -i \\
1 & -1 & 1 & -1 \\
1 & -i & -1 & i
\end{array}\right]=\left[\begin{array}{rccc}
1 & . & 1 & \cdot \\
\cdot & 1 & \cdot & 1 \\
1 & \cdot & -1 & \cdot \\
\cdot & 1 & \cdot & -1
\end{array}\right]\left[\begin{array}{cccc}
1 & \cdot & \cdot & \cdot \\
\cdot & 1 & \cdot & \cdot \\
\cdot & \cdot & 1 & \cdot \\
\cdot & \cdot & \cdot & i
\end{array}\right]\left[\begin{array}{rrrr}
1 & 1 & \cdot & \cdot \\
1 & -1 & \cdot & \cdot \\
\cdot & \cdot & 1 & 1 \\
\cdot & \cdot & 1 & -1
\end{array}\right]\left[\begin{array}{cccc}
1 & \cdot & \cdot & \cdot \\
\cdot & \cdot & 1 & \cdot \\
\cdot & 1 & \cdot & \cdot \\
\cdot & \cdot & \cdot & 1
\end{array}\right]
$$

Representation using matrix algebra

$$
\mathbf{D F T}_{4}=\left(\mathbf{D F T}_{2} \otimes \mathrm{I}_{2}\right) \operatorname{diag}(1,1,1, i)\left(\mathrm{I}_{2} \otimes \mathbf{D F T}_{2}\right) \mathrm{L}_{2}^{4}
$$

Data flow graph (right to left)


## Example Recursive FFT, $\mathbf{n}=16$, radix 4

Kronecker product:
$A=\left[a_{i j}\right]_{i, j}$
$A \otimes B=\left[a_{i j} \cdot B\right]_{i, j}$



General Radix, Recursive Cooley-Tukey FFT
Assume $\mathrm{n}=\mathrm{km}: \quad \mathrm{DFT}_{n}=\left(\mathrm{DFT}_{k} \otimes \mathrm{I}_{m}\right) \mathrm{T}_{m}^{n}\left(\mathrm{I}_{k} \otimes \mathbf{D F T}_{m}\right) \mathrm{L}_{k}^{n}$ permutation matrix diagonal matrix with roots of unity
3 key structures: $\quad \mathrm{I}_{k} \otimes A_{m}, A_{k} \otimes \mathrm{I}_{m}, \mathrm{~L}_{k}^{n}$
$y=\left(\mathrm{I}_{k} \otimes A_{m}\right) x$

for $i=0: k-1$
$y[i m: i m+m-1]=A^{*} x[i m: i m+m-1]$
$y=\left(A_{k} \otimes \mathrm{I}_{m}\right) x$
for $i=0: m-1$
$y[i: m: i+(k-1) m]=A^{*} x[i: m: i+(k-1) m]$
$y=\mathrm{L}_{k}^{n} x \quad$ view x as m x k matrix:

for $i=0: k-1$
for $j=0: m-1$
$y[i m+j]=x[i+k j]$
read at stride $k$, write at stride 1 equivalent: read at stride 1 , write at stride $m$

## Example FFT, $\mathbf{n}=16$ (Recursive, Radix 4)

$\mathrm{DFT}_{4}$ is expanded recursively

 at stride 4

## Recursive Cooley-Tukey FFT

$\mathrm{DFT}_{k m}=\left(\mathrm{DFT}_{k}^{\downarrow} \otimes \mathrm{I}_{m}\right) T_{m}^{k m}\left(\mathrm{I}_{k} \otimes \mathrm{DFT}_{m}\right) L_{k}^{k m} \quad$ decimation-in-time
$\mathrm{DFT}_{k m}=L_{m}^{k m}\left(\mathrm{I}_{k} \otimes \mathrm{DFT}_{m}\right) T_{m}^{k m}\left(\mathrm{DFT}_{k} \otimes \mathrm{I}_{m}\right) \quad$ decimation-in-frequency

For powers of two $n=2^{t}$ sufficient together with base case
$\mathbf{D F T}_{2}=\left[\begin{array}{rr}1 & 1 \\ 1 & -1\end{array}\right]$

Cost:

- $($ complex adds, complex mults $)=\left(n \log _{2}(n), n \log _{2}(n) / 2\right)$ independent of recursion
- (real adds, real mults $) \leq\left(3 n \log _{2}(n), 2 n \log _{2}(n)\right)=5 n \log _{2}(n)$ flops depends on recursion: best is at least radix-8


## Recursive vs. Iterative FFT

Recursive, radix-k Cooley-Tukey FFT

$$
\begin{aligned}
& \mathbf{D F T}_{k m}=\left(\mathbf{D F T}_{k} \otimes \mathbf{I}_{m}\right) T_{m}^{k m}\left(\mathbf{I}_{k} \otimes \mathbf{D F T}_{m}\right) L_{k}^{k m} \\
& \mathbf{D F T}_{k m}=L_{m}^{k m}\left(\mathbf{I}_{k} \otimes \mathbf{D F T}_{m}\right) T_{m}^{k m}\left(\mathbf{D F T}_{k} \otimes \mathbf{I}_{m}\right)
\end{aligned}
$$

Iterative, radix 2, decimation-in-time/decimation-in-frequency

$$
\begin{aligned}
& \mathrm{DFT}_{2^{t}}=\left(\prod_{j=1}^{t}\left(\mathrm{I}_{2^{j-1}} \otimes \mathrm{DFT}_{2} \otimes \mathrm{I}_{2^{t-j}}\right) \cdot\left(\mathrm{I}_{2^{j-1}} \otimes T_{2^{t-j}}^{2^{t-j+1}}\right)\right) \cdot R_{2^{t}} \\
& \mathrm{DFT}_{2^{t}}=R_{2^{t}} \cdot\left(\prod_{j=1}^{t}\left(\mathrm{I}_{2^{t-j}} \otimes T_{2^{j-1}}^{2^{j}}\right) \cdot\left(\mathrm{I}_{2^{t-j}} \otimes \mathbf{D F T}_{2} \otimes \mathrm{I}_{2^{j-1}}\right)\right)
\end{aligned}
$$



## Recursive vs. Iterative

Iterative FFT computes in stages of butterflies =
$\log _{2}(n)$ passes through the data
Recursive FFT reduces passes through data $=$ better locality

Same computation graph but different topological sorting

Rough analogy:

## MMM

DFT
Triple loop
Iterative FFT
Blocked
Recursive FFT

The FFT Is Very Malleable

## Iterative FFT, Radix 2


$\left(\left(I_{1} \otimes \mathrm{DFT}_{2} \otimes I_{8}\right) D_{0}^{16}\right)\left(\left(I_{2} \otimes \mathrm{DFT}_{2} \otimes I_{4}\right) D_{1}^{16}\right)\left(\left(I_{4} \otimes \mathrm{DFT}_{2} \otimes I_{2}\right) D_{2}^{16}\right)\left(\left(I_{8} \otimes \mathrm{DFT}_{2} \otimes I_{1}\right) D_{3}^{16}\right) R_{2}^{16}$

## Pease FFT, Radix 2


$\left(L_{2}^{16}\left(I_{8} \otimes \mathrm{DFT}_{2}\right) D_{0}^{16}\right)\left(L_{2}^{16}\left(I_{8} \otimes \mathrm{DFT}_{2}\right) D_{1}^{16}\right)\left(L_{2}^{16}\left(I_{8} \otimes \mathrm{DFT}_{2}\right) D_{2}^{16}\right)\left(L_{2}^{16}\left(I_{8} \otimes \mathrm{DFT}_{2}\right) D_{3}^{16}\right) R_{2}^{16}$

## Stockham FFT, Radix 2


$\left(\left(\mathrm{DFT}_{2} \otimes I_{8}\right) D_{0}^{16}\left(L_{2}^{2} \otimes I_{8}\right)\right)\left(\left(\mathrm{DFT}_{2} \otimes I_{8}\right) D_{1}^{16}\left(L_{2}^{4} \otimes I_{4}\right)\right)\left(\left(\mathrm{DFT}_{2} \otimes I_{8}\right) D_{2}^{16}\left(L_{2}^{8} \otimes I_{2}\right)\right)\left(\left(\mathrm{DFT}_{2} \otimes I_{8}\right) D_{3}^{16}\left(L_{2}^{16} \otimes I_{1}\right)\right)$

## Six-Step FFT


$L_{4}^{16}\left(I_{4} \otimes\left(\left(\mathrm{DFT}_{2} \otimes I_{2}\right) T_{2}^{4}\left(I_{2} \otimes \mathrm{DFT}_{2}\right) L_{2}^{4}\right)\right) L_{4}^{16} T_{4}^{16}\left(I_{4} \otimes\left(\left(\mathrm{DFT}_{2} \otimes I_{2}\right) T_{2}^{4}\left(I_{2} \otimes \mathrm{DFT}_{2}\right) L_{2}^{4}\right)\right) L_{4}^{16}$

## Multi-Core FFT


$\left(L_{4}^{8} \otimes I_{2}\right)\left(I_{2} \otimes\left(\left(\mathrm{DFT}_{2} \otimes I_{2}\right) T_{2}^{4}\left(I_{2} \otimes \mathrm{DFT}_{2}\right) L_{2}^{4}\right) \otimes I_{2}\right)\left(L_{2}^{8} \otimes I_{2}\right) T_{4}^{16}\left(I_{2} \otimes\left(I_{2} \otimes\left(\mathrm{DFT}_{2} \otimes I_{2}\right) T_{2}^{4}\left(I_{2} \otimes \mathrm{DFT}_{2}\right)\right) R_{2}^{8}\right)\left(L_{2}^{8} \otimes I_{2}\right)$

## Transform Algorithms

```
            DFT
        |}|\begin{array}{l}{\mp@subsup{\mathbf{RDFT}}{n}{\prime}}\\{\mp@subsup{\mathbf{RDFT}}{n}{\prime}}\\{\mp@subsup{\textrm{DHT}}{n}{\prime}}\\{\mp@subsup{\mathbf{DHT}}{n}{\prime}}\end{array}|\mp@code{( (P/2,m
```



```
    RDFT-3}n->(\mp@subsup{Q}{k/2,m}{\top}\otimes\mp@subsup{I}{2}{})(\mp@subsup{I}{k}{}\mp@subsup{\otimes}{i}{}\mp@subsup{\textrm{rDFT}}{2m}{\prime})(i+1/2)/k))(RDFT- 3 < \otimesIm), k even
        DCT-2 2n}->\mp@subsup{P}{k/2,2m}{\top}(\mathrm{ DCT- 22m K
        DCT-3 }\mp@subsup{n}{n}{}->\mathrm{ DCT-2 - }\mp@subsup{n}{n}{\top
        DCT-4}\mp@subsup{|}{n}{}->\mp@subsup{Q}{k/2,2m}{\top}(\mp@subsup{I}{k/2}{}\otimes\mp@subsup{N}{2m}{}\mathrm{ RDFT-32m
```




```
            DFT
    DCT-3}\mp@subsup{\mathbf{3}}{n}{}->(\mp@subsup{\textrm{I}}{m}{}\oplus\mp@subsup{\textrm{J}}{m}{})\mp@subsup{\llcorner}{m}{n}(\mathrm{ DCT-3
\[
\cdot\left(F_{2} \otimes \mathrm{I}_{m}\right)\left[\begin{array}{cc}
\mathrm{I}_{m} & 0 \oplus-\mathrm{J}_{m-1} \\
\frac{1}{\sqrt{2}}\left(\mathrm{I}_{1} \oplus 2 \mathbf{I}_{m}\right)
\end{array}\right], \quad n=2 m
\]
DCT-4 \(n \rightarrow S_{n}\) DCT- \(2_{n}\) diag \(_{0 \leq k<n}(1 /(2 \cos ((2 k+1) \pi / 4 n)))\)
\(\mathrm{IMDCT}_{2 m} \rightarrow\left(\mathrm{~J}_{m} \oplus \mathrm{I}_{m} \oplus \mathrm{I}_{m} \oplus \mathrm{~J}_{m}\right)\left(\left(\left[\begin{array}{r}1 \\ -1\end{array}\right] \otimes \mathrm{I}_{m}\right) \oplus\left(\left[\begin{array}{c}-1 \\ -1\end{array}\right] \otimes \mathbf{I}_{m}\right)\right) \mathrm{J}_{2 m} \mathbf{D C T}-\mathbf{4}_{2 m}\)
\(\mathrm{WHT}_{2^{k}} \rightarrow \prod_{i=1}^{t}\left(\mathrm{I}_{2^{k_{1}+\cdots+k_{i-1}}} \otimes \mathrm{WHT}_{2^{k_{i}}} \otimes \mathrm{I}_{2^{k_{i+1}}+\cdots+k_{t}}\right), \quad k=k_{1}+\cdots+k_{t}\)
\(\mathrm{DFT}_{2} \rightarrow \mathrm{~F}_{2}\)
DCT- \(\mathbf{2}_{2} \rightarrow \operatorname{diag}(1,1 / \sqrt{2}) F_{2}\)
DCT-4 \({ }_{2} \rightarrow \mathrm{~J}_{2} \mathrm{R}_{13 \pi / 8}\)
```


## Complexity of the DFT

Measure: $L_{c}, 2 \leq c$

- Complex adds count 1
- Complex mults by a constant a with $|a|<c$ counts 1
- $L_{2}$ is strictest, $L_{\infty}$ the loosest (and most natural)

Upper bounds:

- $n=2^{k}: \quad L_{2}\left(D F T_{n}\right) \leq 3 / 2 n \log _{2}(n) \quad$ (using Cooley-Tukey FFT)
- General n: $L_{2}\left(D F T_{n}\right) \leq 8 n \log _{2}(n) \quad$ (needs Bluestein FFT)

Lower bound:

- Theorem by Morgenstern: If $c<\infty$, then $L_{c}\left(D F T_{n}\right) \geq 1 / 2 n \log _{c}(n)$
- Implies: in the measure $L_{c}$ the DFT is $\Theta(n \log (n))$


## Lowest Known FFT Cost (Powers of 2)

A modified split-radix FFT with fewer arithmetic operations, Johnson and Frigo, IEEE Trans. Signal Processing 55(1), pp. 111-119, 2007

Number of flops ( $\mathrm{n}=2^{\mathrm{k}}$ ):

$$
\frac{34}{9} n \log _{2}(n)-\frac{124}{27} n-2 \log _{2}(n)-\frac{2}{9}(-1)^{\log _{2}(n)} \log _{2}(n)+\frac{16}{27}(-1)^{\log _{2}(n)}+8
$$

## History of FFTs

The advent of digital signal processing is often attributed to the FFT
(Cooley-Tukey 1965)
History:

- Around 1805: FFT discovered by Gauss [1] (Fourier publishes the concept of Fourier analysis in 1807!)
- 1965: Rediscovered by Cooley-Tukey


## Carl-Friedrich Gauss



1777-1855

Contender for the greatest mathematician of all times
Some contributions: Modular arithmetic, least square analysis, normal distribution, fundamental theorem of algebra, Gauss elimination, Gauss quadrature, Gauss-Seidel, non-Euclidean geometry, ...

