

How to Write Fast Numerical Code

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Lecture: Discrete Fourier transform, fast Fourier transforms

Instructor: Markus Püschel

TA: Gagandeep Singh, Daniele Spampinato, Alen Stojanov



Eidgenössische Technische Hochschule Zürich
Swiss Federal Institute of Technology Zurich

Linear Transforms

- **Overview: Transforms and algorithms**
- **Discrete Fourier transform**
- **Fast Fourier transforms**
- **After that:**
 - Optimized implementation and autotuning (FFTW)
 - Automatic program synthesis (Spiral)

Blackboard

- Linear Transforms
- Discrete Fourier transform (DFT)
- Transform algorithms
- Fast Fourier transform, size 4

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FFT References

- **Complexity:** *Bürgisser, Clausen, Shokrollahi, Algebraic Complexity Theory, Springer, 1997*
- **History:** *Heideman, Johnson, Burrus: Gauss and the History of the Fast Fourier Transform, Arch. Hist. Sc. 34(3) 1985*
- **FFTs:**
 - *Cooley and Tukey, An algorithm for the machine calculation of complex Fourier series," Math. of Computation, vol. 19, pp. 297–301, 1965*
 - *Nussbaumer, Fast Fourier Transform and Convolution Algorithms, 2nd ed., Springer, 1982*
 - *van Loan, Computational Frameworks for the Fast Fourier Transform, SIAM, 1992*
 - *Tolimieri, An, Lu, Algorithms for Discrete Fourier Transforms and Convolution, Springer, 2nd edition, 1997*
 - *Franchetti, Püschel, Voronenko, Chellappa and Moura, Discrete Fourier Transform on Multicore, IEEE Signal Processing Magazine, special issue on "Signal Processing on Platforms with Multiple Cores", Vol. 26, No. 6, pp. 90-102, 2009*

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Linear Transforms

- Very important class of functions: signal processing, scientific computing, ...
- **Mathematically:** Change of basis = Multiplication by a fixed matrix T

$$\begin{pmatrix} y_0 \\ y_1 \\ \vdots \\ y_{n-1} \end{pmatrix} = y = Tx \leftarrow \boxed{T} \leftarrow x = \begin{pmatrix} x_0 \\ x_1 \\ \vdots \\ x_{n-1} \end{pmatrix}$$

$T = [t_{k,\ell}]_{0 \leq k, \ell < n}$

Output **Input**

- Equivalent definition: Summation form

$$y_k = \sum_{\ell=0}^{n-1} t_{k,\ell} x_{\ell}, \quad 0 \leq k < n$$

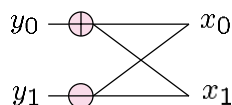
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Smallest Relevant Example: DFT, Size 2

Transform (matrix): $T = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$

Computation: $y = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} x$ or $\begin{aligned} y_0 &= x_0 + x_1 \\ y_1 &= x_0 - x_1 \end{aligned}$

As graph (direct acyclic graph or DAG):



called a butterfly



http://charlottismartyns.blogspot.com/2011_02_01_archive.html

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Transforms: Examples

- A few dozen transforms are relevant
- Some examples

$$\begin{aligned}
 \text{DFT}_n &= [e^{-2k\ell\pi i/n}]_{0 \leq k, \ell < n} \\
 \text{RDFT}_n &= [r_{k\ell}]_{0 \leq k, \ell < n}, \quad r_{k\ell} = \begin{cases} \cos \frac{2\pi k\ell}{n}, & k \leq \lfloor \frac{n}{2} \rfloor \\ -\sin \frac{2\pi k\ell}{n}, & k > \lfloor \frac{n}{2} \rfloor \end{cases} \quad \text{universal tool} \\
 \text{DHT} &= [\cos(2k\ell\pi/n) + \sin(2k\ell\pi/n)]_{0 \leq k, \ell < n} \\
 \text{WHT}_n &= \begin{bmatrix} \text{WHT}_{n/2} & \text{WHT}_{n/2} \\ \text{WHT}_{n/2} & -\text{WHT}_{n/2} \end{bmatrix}, \quad \text{WHT}_2 = \text{DFT}_2 \\
 \text{IMDCT}_n &= [\cos((2k+1)(2\ell+1+n)\pi/4n)]_{0 \leq k < 2n, 0 \leq \ell < n} \quad \text{MPEG} \\
 \text{DCT-2}_n &= [\cos(k(2\ell+1)\pi/2n)]_{0 \leq k, \ell < n} \quad \text{JPEG} \\
 \text{DCT-3}_n &= \text{DCT-2}_n^T \quad (\text{transpose}) \\
 \text{DCT-4}_n &= [\cos((2k+1)(2\ell+1)\pi/4n)]_{0 \leq k, \ell < n}
 \end{aligned}$$

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Linear Transforms: DFT

$$\begin{pmatrix} y_0 \\ y_1 \\ \vdots \\ y_{n-1} \end{pmatrix} = y = Tx \quad \leftarrow \quad \boxed{T} \quad \leftarrow \quad x = \begin{pmatrix} x_0 \\ x_1 \\ \vdots \\ x_{n-1} \end{pmatrix}$$

$T = [t_{k,\ell}]_{0 \leq k, \ell < n}$

Output **Input**

Example: $T = \text{DFT}_n = [e^{-2k\ell\pi i/n}]_{0 \leq k, \ell < n}$

$$= [\omega_n^{k\ell}]_{0 \leq k, \ell < n}, \quad \omega_n = e^{-2\pi i/n}$$

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Algorithms: Example FFT, $n = 4$

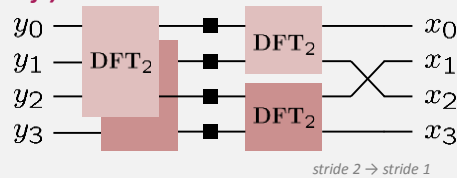
Fast Fourier transform (FFT)

$$\begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & i & -1 & -i \\ 1 & -1 & 1 & -1 \\ 1 & -i & -1 & i \end{bmatrix} x = \begin{bmatrix} 1 & \cdot & 1 & \cdot \\ \cdot & 1 & \cdot & 1 \\ 1 & \cdot & -1 & \cdot \\ \cdot & 1 & \cdot & -1 \end{bmatrix} \begin{bmatrix} 1 & \cdot & \cdot & \cdot \\ \cdot & 1 & \cdot & \cdot \\ \cdot & \cdot & 1 & \cdot \\ \cdot & \cdot & \cdot & i \end{bmatrix} \begin{bmatrix} 1 & 1 & \cdot & \cdot \\ 1 & -1 & \cdot & \cdot \\ \cdot & \cdot & 1 & 1 \\ \cdot & \cdot & 1 & -1 \end{bmatrix} \begin{bmatrix} 1 & \cdot & \cdot & \cdot \\ \cdot & \cdot & 1 & \cdot \\ \cdot & 1 & \cdot & \cdot \\ \cdot & \cdot & \cdot & 1 \end{bmatrix} x$$

Representation using matrix algebra

$$\text{DFT}_4 = (\text{DFT}_2 \otimes \text{I}_2) \text{diag}(1, 1, 1, i)(\text{I}_2 \otimes \text{DFT}_2) \text{L}_2^4$$

Data flow graph (right to left)



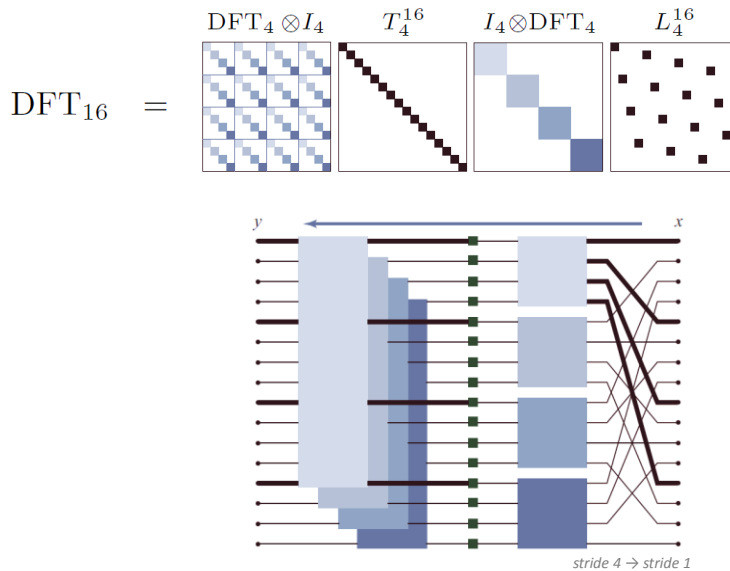
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Cooley-Tukey FFT (Recursive, General-Radix)

- Blackboard
- Kronecker products
- Stride permutations

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Example FFT, $n = 16$ (Recursive, Radix 4)



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Recursive Cooley-Tukey FFT

Diagram illustrating the Recursive Cooley-Tukey FFT structure. The radix is indicated by a red arrow pointing to DFT_k in the equation:

$$\text{DFT}_{km} = (\text{DFT}_k \otimes I_m) T_m^{km} (I_k \otimes \text{DFT}_m) L_k^{km} \quad \text{decimation-in-time}$$

$$\text{DFT}_{km} = L_m^{km} (I_k \otimes \text{DFT}_m) T_m^{km} (\text{DFT}_k \otimes I_m) \quad \text{decimation-in-frequency}$$

- For powers of two $n = 2^l$ sufficient together with base case

$$\text{DFT}_2 = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$$

- Cost:

- (complex adds, complex mults) = $(n \log_2(n), n \log_2(n)/2)$
independent of recursion
- (real adds, real mults) $\leq (2n \log_2(n), 3n \log_2(n)) = 5n \log_2(n)$ flops
depends on recursion: best is at least radix-8

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Recursive vs. Iterative FFT

■ Recursive, radix-k Cooley-Tukey FFT

$$\text{DFT}_{km} = (\text{DFT}_k \otimes \text{I}_m) T_m^{km} (\text{I}_k \otimes \text{DFT}_m) L_k^{km}$$

$$\text{DFT}_{km} = L_m^{km} (\text{I}_k \otimes \text{DFT}_m) T_m^{km} (\text{DFT}_k \otimes \text{I}_m)$$

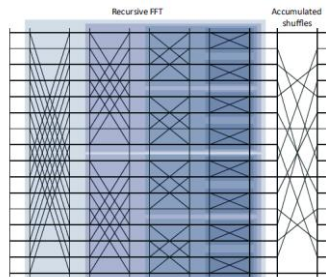
■ Iterative, radix 2, decimation-in-time/decimation-in-frequency

$$\text{DFT}_{2^t} = \left(\prod_{j=1}^t (\text{I}_{2^{j-1}} \otimes \text{DFT}_2 \otimes \text{I}_{2^{t-j}}) \cdot (\text{I}_{2^{j-1}} \otimes T_{2^{t-j}}^{2^{t-j+1}}) \right) \cdot R_{2^t}$$

$$\text{DFT}_{2^t} = R_{2^t} \cdot \left(\prod_{j=1}^t (\text{I}_{2^{t-j}} \otimes T_{2^{j-1}}^{2^j}) \cdot (\text{I}_{2^{t-j}} \otimes \text{DFT}_2 \otimes \text{I}_{2^{j-1}}) \right)$$

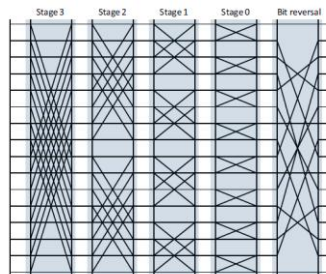
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Radix 2, recursive



$$(\text{DFT}_2 \otimes \text{I}_8) T_8^{16} \left(\text{I}_2 \otimes \left((\text{DFT}_2 \otimes \text{I}_4) T_4^8 \left(\text{I}_2 \otimes \left((\text{DFT}_2 \otimes \text{I}_2) T_2^4 (\text{I}_2 \otimes \text{DFT}_2) L_2^8 \right) \right) \right) L_2^{16} \right)$$

Radix 2, iterative



$$\left((\text{I}_1 \otimes \text{DFT}_2 \otimes \text{I}_8) D_0^{16} \right) \left((\text{I}_2 \otimes \text{DFT}_2 \otimes \text{I}_4) D_1^{16} \right) \left((\text{I}_4 \otimes \text{DFT}_2 \otimes \text{I}_2) D_2^{16} \right) \left((\text{I}_8 \otimes \text{DFT}_2 \otimes \text{I}_1) D_3^{16} \right) R_2^{16}$$

Recursive vs. Iterative

- **Iterative FFT computes in stages of butterflies =**
 $\log_2(n)$ passes through the data
- **Recursive FFT reduces passes through data =**
better locality
- **Same computation graph but different topological sorting**
- **Rough analogy:**

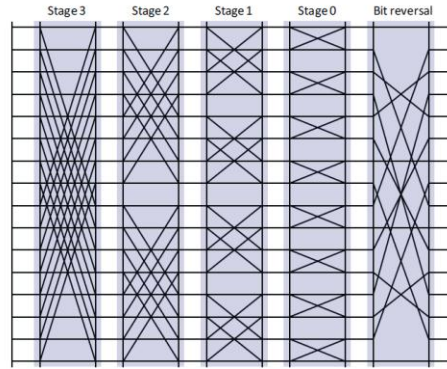
MMM	DFT
Triple loop	Iterative FFT
Blocked	Recursive FFT

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The FFT Is Very Malleable

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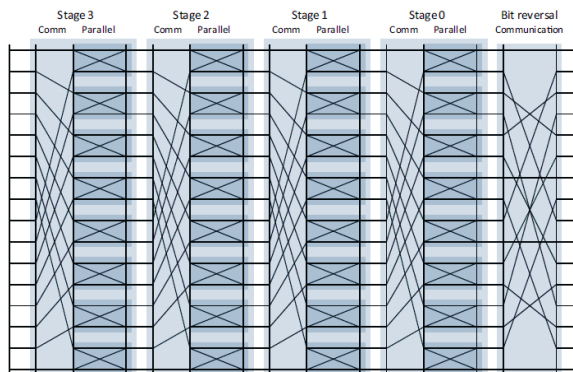
Iterative FFT, Radix 2



$$\left((I_1 \otimes \text{DFT}_2 \otimes I_8) D_0^{16} \right) \left((I_2 \otimes \text{DFT}_2 \otimes I_4) D_1^{16} \right) \left((I_4 \otimes \text{DFT}_2 \otimes I_2) D_2^{16} \right) \left((I_8 \otimes \text{DFT}_2 \otimes I_1) D_3^{16} \right) R_2^{16}$$

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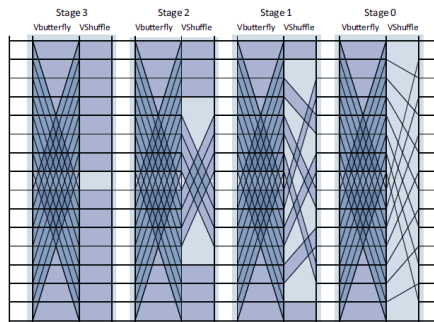
Pease FFT, Radix 2



$$\left(L_2^{16} (I_8 \otimes \text{DFT}_2) D_0^{16} \right) \left(L_2^{16} (I_8 \otimes \text{DFT}_2) D_1^{16} \right) \left(L_2^{16} (I_8 \otimes \text{DFT}_2) D_2^{16} \right) \left(L_2^{16} (I_8 \otimes \text{DFT}_2) D_3^{16} \right) R_2^{16}$$

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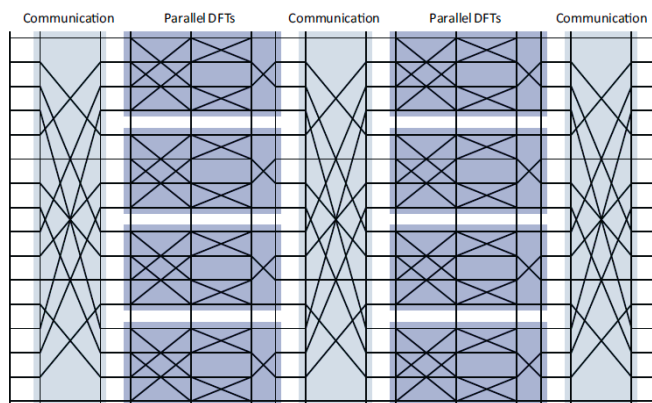
Stockham FFT, Radix 2



$$\left(\left(\text{DFT}_2 \otimes I_8 \right) D_0^{16} \left(L_2^2 \otimes I_8 \right) \right) \left(\left(\text{DFT}_2 \otimes I_8 \right) D_1^{16} \left(L_2^4 \otimes I_4 \right) \right) \left(\left(\text{DFT}_2 \otimes I_8 \right) D_2^{16} \left(L_2^8 \otimes I_2 \right) \right) \left(\left(\text{DFT}_2 \otimes I_8 \right) D_3^{16} \left(L_2^{16} \otimes I_1 \right) \right)$$

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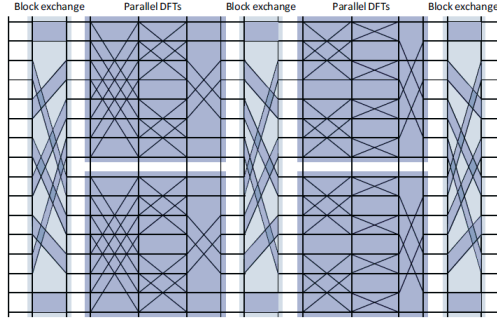
Six-Step FFT



$$L_4^{16} \left(I_4 \otimes \left(\left(\text{DFT}_2 \otimes I_2 \right) T_2^4 \left(I_2 \otimes \text{DFT}_2 \right) L_2^4 \right) \right) L_4^{16} T_4^{16} \left(I_4 \otimes \left(\left(\text{DFT}_2 \otimes I_2 \right) T_2^4 \left(I_2 \otimes \text{DFT}_2 \right) L_2^4 \right) \right) L_4^{16}$$

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Multi-Core FFT



$$(L_4^8 \otimes I_2) \left(I_2 \otimes \left((DFT_2 \otimes I_2) T_2^4 (I_2 \otimes DFT_2) L_2^4 \right) \otimes I_2 \right) (L_2^8 \otimes I_2) T_4^{16} \left(I_2 \otimes \left(I_2 \otimes (DFT_2 \otimes I_2) T_2^4 (I_2 \otimes DFT_2) \right) R_2^8 \right) (L_2^8 \otimes I_2)$$

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Transform Algorithms

$$\begin{aligned} DFT_n &\rightarrow P_{k/2,2m}^\top \left(DFT_{2m} \oplus (I_{k/2-1} \otimes C_{2m} rDFT_{2m}(i/k)) \right) (RDFT_k' \otimes I_m), \quad k \text{ even,} \\ \begin{bmatrix} RDFT_n \\ RDFT_n' \\ DHT_n \\ DHT_n' \end{bmatrix} &\rightarrow (P_{k/2,2m}^\top \otimes I_2) \left(\begin{bmatrix} RDFT_{2m}' \\ RDFT_{2m}' \\ DHT_{2m}' \\ DHT_{2m}' \end{bmatrix} \oplus \left(I_{k/2-1} \otimes D_{2m} \begin{bmatrix} rDFT_{2m}(i/k) \\ rDFT_{2m}(i/k) \\ rDHT_{2m}(i/k) \\ rDHT_{2m}(i/k) \end{bmatrix} \right) \right) \begin{bmatrix} RDFT_k' \\ RDFT_k' \\ DHT_k' \\ DHT_k' \end{bmatrix} \otimes I_m, \quad k \text{ even,} \\ \begin{bmatrix} rDFT_{2n}(u) \\ rDHT_{2n}(u) \end{bmatrix} &\rightarrow L_m^{2n} \left(I_k \otimes \begin{bmatrix} rDFT_{2m}((i+u)/k) \\ rDHT_{2m}((i+u)/k) \end{bmatrix} \right) \begin{bmatrix} rDFT_{2k}(u) \\ rDHT_{2k}(u) \end{bmatrix} \otimes I_m, \\ RDFT\text{-}3_n &\rightarrow (Q_{k/2,2m}^\top \otimes I_2) (I_k \otimes rDFT_{2m}(i+1/2)/k) (RDFT\text{-}3_k \otimes I_m), \quad k \text{ even,} \\ DCT\text{-}2_n &\rightarrow P_{k/2,2m}^\top (DCT\text{-}2_{2m} K_{2m}^{2m} \oplus (I_{k/2-1} \otimes N_{2m} RDFT\text{-}3_{2m}^\top)) B_n(L_{k/2}^{n/2} \otimes I_2) (I_m \otimes RDFT_k' Q_{m/2,k}), \\ DCT\text{-}3_n &\rightarrow DCT\text{-}2_n^\top, \\ DCT\text{-}4_n &\rightarrow Q_{k/2,2m}^\top (I_{k/2} \otimes N_{2m} RDFT\text{-}3_{2m}^\top) B_n'(L_{k/2}^{n/2} \otimes I_2) (I_m \otimes RDFT\text{-}3_k) Q_{m/2,k}, \\ DFT_n &\rightarrow (DFT_k \otimes I_m) T_m^n (I_k \otimes DFT_m) L_m^n, \quad n = km \quad \text{Cooley-Tukey FFT} \\ DFT_n &\rightarrow P_n (DFT_k \otimes DFT_m) Q_n, \quad n = km, \quad \gcd(k, m) = 1 \quad \text{Prime-factor FFT} \\ DFT_p &\rightarrow R_p^\top (I_1 \oplus DFT_{p-1}) D_p (I_1 \oplus DFT_{p-1}) R_p, \quad p \text{ prime} \quad \text{Rader FFT} \\ DCT\text{-}3_n &\rightarrow (I_m \oplus J_m) L_m^n (DCT\text{-}3_m(1/4) \oplus DCT\text{-}3_m(3/4)) \\ &\quad \cdot (F_2 \otimes I_m) \begin{bmatrix} I_m & 0 & -J_{m-1} \\ 0 & I_1 & 2I_m \end{bmatrix}, \quad n = 2m \\ DCT\text{-}4_n &\rightarrow S_n DCT\text{-}2_n \text{diag}_{0 \leq k < n} (1/(2 \cos((2k+1)\pi/4n))) \\ IMDCT_{2m} &\rightarrow (J_m \otimes I_m \oplus I_m \oplus J_m) \left(\begin{bmatrix} 1 \\ 1 \\ -1 \end{bmatrix} \otimes I_m \right) \oplus \left(\begin{bmatrix} -1 \\ -1 \end{bmatrix} \otimes I_m \right) J_{2m} DCT\text{-}4_{2m} \\ WHT_{2^k} &\rightarrow \prod_{i=1}^k (I_{2^{k_1+\dots+k_{i-1}}} \otimes WHT_{2^{k_i}} \otimes I_{2^{k_{i+1}+\dots+k_l}}), \quad k = k_1 + \dots + k_l \\ DFT_2 &\rightarrow F_2 \\ DCT\text{-}2_2 &\rightarrow \text{diag}(1, 1/\sqrt{2}) F_2 \\ DCT\text{-}4_2 &\rightarrow J_2 R_{13\pi/8} \end{aligned}$$

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Complexity of the DFT

■ Measure: L_c , $2 \leq c$

- Complex adds count 1
- Complex mult by a constant a with $|a| < c$ counts 1
- L_2 is strictest, L_∞ the loosest (and most natural)

■ Upper bounds:

- $n = 2^k$: $L_2(\text{DFT}_n) \leq 3/2 n \log_2(n)$ *(using Cooley-Tukey FFT)*
- General n : $L_2(\text{DFT}_n) \leq 8 n \log_2(n)$ *(needs Bluestein FFT)*

■ Lower bound:

- Theorem by Morgenstern: If $c < \infty$, then $L_c(\text{DFT}_n) \geq \frac{1}{2} n \log_c(n)$
- Implies: in the measure L_c , the DFT is $\Theta(n \log(n))$

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History of FFTs

■ The advent of digital signal processing is often attributed to the FFT (Cooley-Tukey 1965)

■ History:

- Around 1805: FFT discovered by Gauss [1]
(Fourier publishes the concept of Fourier analysis in 1807!)
- 1965: Rediscovered by Cooley-Tukey

[1]: Heideman, Johnson, Burrus: "Gauss and the History of the Fast Fourier Transform" Arch. Hist. Sc. 34(3) 1985²⁴

Carl-Friedrich Gauss



1777 - 1855

- **Contender for the greatest mathematician of all times**
- **Some contributions:** Modular arithmetic, least square analysis, normal distribution, fundamental theorem of algebra, Gauss elimination, Gauss quadrature, Gauss-Seidel, non-Euclidean geometry, ...

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